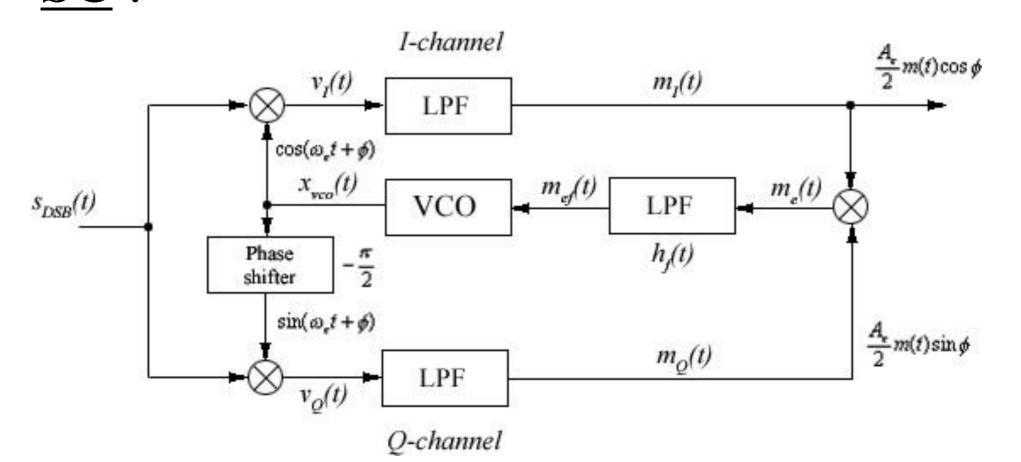


DSBSC RECEIVER CIRCUIT COSTAS LOOP

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<u>Coherent Costas loop receiver for DSB-</u> <u>SC</u>:



I-channel:

After down converwsion,

$$v_{I}(t) = A_{c}m(t)\cos(\omega_{c}t)\cdot\cos(\omega_{c}t+\phi)$$
$$= \frac{A_{c}}{2}m(t)\left[\cos(2\omega_{c}t+\phi)+\cos\phi\right]$$

At the output of the lowpass filter, with |H(0)| = 1,

$$m_I(t) = \frac{A_c}{2} \cos \phi \cdot m(t)$$

Q-channel:

$$v_Q(t) = \frac{A_c}{2} m(t) \left[\sin\left(2\omega_c t + \phi\right) + \sin\phi \right]$$
$$m_Q(t) = \frac{A_c}{2} \sin\phi \cdot m(t)$$

Feedback path:

At the output of the multiplier,

$$m_e(t) = \frac{A_c^2}{4} m^2(t) \sin \phi \cos \phi$$
$$= \frac{A_c^2}{8} m^2(t) \sin 2\phi$$

At the output of the lowpass filter,

$$m_{ef}(t) = \int_{-\infty}^{\infty} m_e(\tau) h_f(t-\tau) d\tau$$

The purpose of $h_f(t)$ is to smooth out fast time variations of $m_e(t)$

The output of the VCO is described by

$$x_{VCO}(t) = \cos(\omega_c t + \phi)$$

Where ω_c is the VCO's reference frequency and $\phi(t) = k_v \int_0^t m_{ef}(\tau) d\tau$, is the residual phase angle due to the tracking error. The constant k_v is the frequency sensitivity of the VCO in rad/s/volt

The instantaneous frequency in radians/sec of the VCO's output is given by

$$\frac{d\left[\omega_{c}t+\phi(t)\right]}{dt}=\omega_{c}+k_{v}m_{ef}(t),$$

Clearly, if $\phi(t)$ were small, then the instantaneous frequency would be close to ω_c and the output of the I-path would also be proportional to m(t).



thank to the test of t