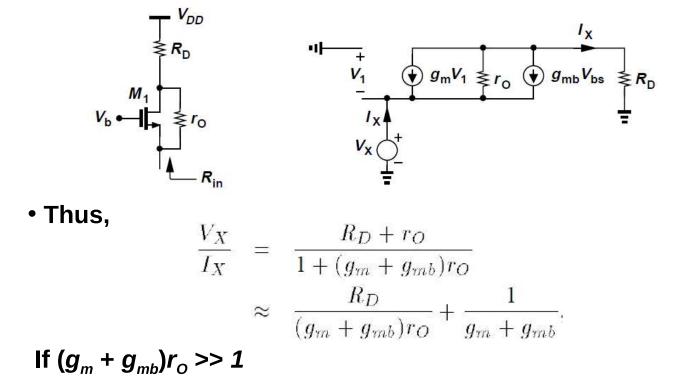


 From the small-signal equivalent circuit for finding input impedance, we have

$$V_1 = -V_X$$

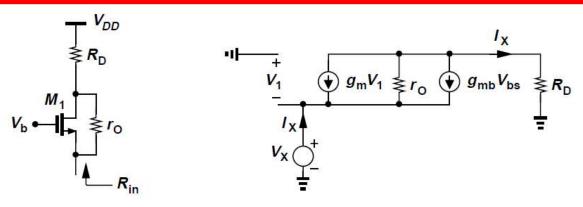
- The current through  $r_O$  is equal to  $I_X + g_m V_1 + g_{mb} V_1 = I_X (g_m + g_{mb}) V_X$
- Voltages across  $r_o$  and  $R_D$  can be added and equated to

$$R_D I_X + r_O [I_X - (g_m + g_{mb})V_X] = V_X$$



- The drain impedance is divided by  $(g_m + g_{mb})r_o$  when seen at the source
- Important in short-channel devices because of their Copyright © 2017 MOW HILL IN SIC REPRODUCTION OF DISTRIBUTION WITHOUT THE PRIOR WRITTEN CONSENT OF MCGraw-Hill Education.

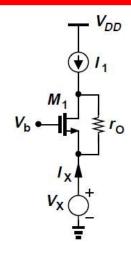
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• Suppose  $R_D = 0$ , then

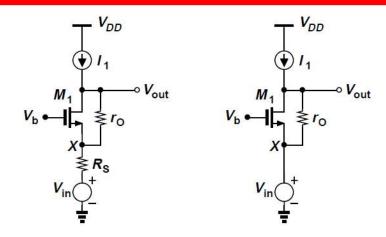
$$\frac{V_X}{I_X} = \frac{r_O}{1 + (g_m + g_{mb})r_O} \\
= \frac{1}{\frac{1}{r_O} + g_m + g_{mb}},$$

• This is the impedance seen at the source of a source follower, a predictable result since with  $R_D = 0$  the circuit configuration is the same as a source follower



- If  $R_D$  is replaced with an ideal current source, earlier result predicts that input impedance approaches infinity
- Total current through the transistor is fixed and is equal to  $I_1$
- Therefore, a change in the source potential cannot change the device current, and hence  $I_x = 0$
- The input impedance of a CG stage is relatively low *only* if the load impedance connected to the drain is small

### **Common-Gate Stage**

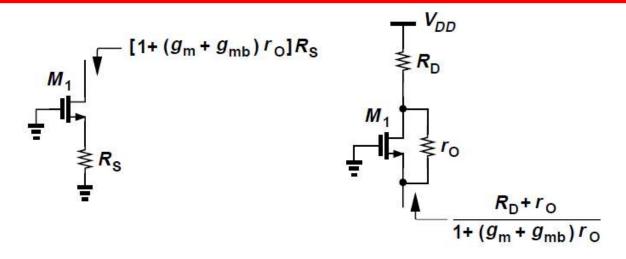


• In a CG stage with a current source load, substituting  $R_D = \infty$  in the voltage gain equation, we get

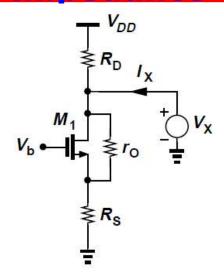
$$A_v = (g_m + g_{mb})r_O + 1$$

- $\bullet$  Gain does not depend on  $R_s$
- From the foregoing discussion, if  $R_D \to \infty$ , so does the impedance seen at the source of  $M_1$ , and the small-signal voltage at node X becomes equal to  $V_{in}$

# **Common-Gate Stage**



- In a degenerated CS stage, we loosely say that a transistor transforms its source resistance *up*
- In a CG stage, the transistor transforms its drain resistance *down*
- The MOS transistor can thus be viewed as an resistance transformer



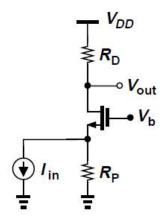
 From the above small-signal equivalent circuit, we can find output impedance as

$$R_{out} = \{[1 + (g_m + g_{mb})r_O]R_S + r_O\} ||R_D$$

Result is similar to that obtained for a degenerated CS stage

### **Common-Gate Stage**

• Input signal of a common-gate stage may be a current rather than a voltage as shown below



- Input current source exhibits output impedance of  $R_P$
- To find the "gain"  $V_{out}II_{in}$ , replace  $I_{in}$  and  $R_P$  with a Thevenin equivalent and use derived result to write

Thevenin equivalent and use derived result to write 
$$\frac{V_{out}}{I_{in}} = \frac{(g_m + g_{mb})r_O + 1}{r_O + (g_m + g_{mb})r_OR_P + R_P + R_D}R_DR_P$$

$$R_{out} = \{ [1 + (g_m + g_{mb})r_O]R_P + r_O \} ||R_D|$$

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